Overview Electrical Machines and Drives

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- 11-9 1.2-3: Magnetic circuits, Principles
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- 18-9 4.3-4.7: DC machines and drives
- 21-9 5.2-5.6: IM introduction, IM principles
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- 28-9 5.8-5.10: IM equivalent circuits and characteristics
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Induction machines

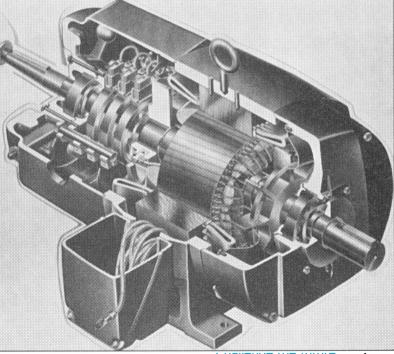
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Recapitulation

- Why is the induction machine the workhorse of industry?
- How does the stator of an induction machine create a rotating magnetic field?
- Why does the induction machine develop torque?



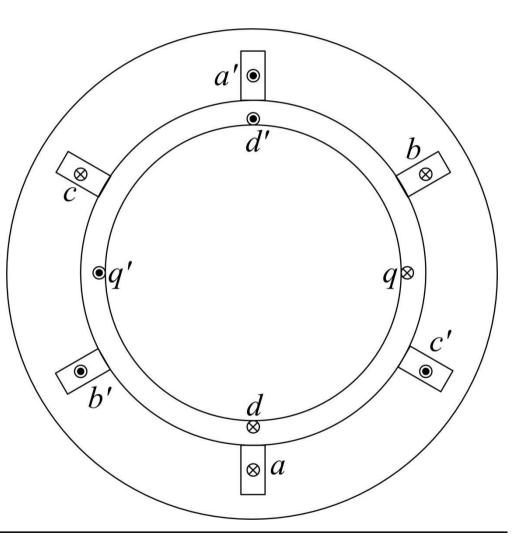




Why does an IM work?

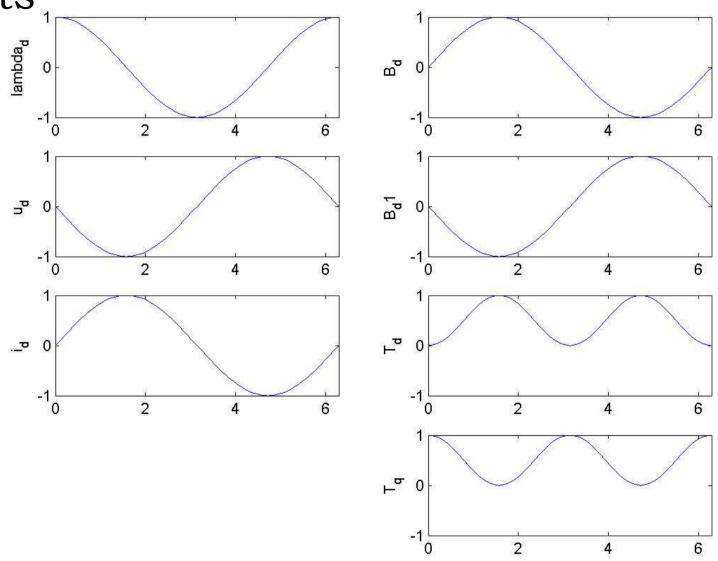
 $B_{s}(\theta,t) = \hat{B}\cos(\theta - \omega t)$

- 1 Sketch the flux linkage of dd'.
- 2 Sketch induced voltage in dd'.
- The rotor turns *dd* and *qq* are short-circuited via a large resistance R_r .
- 3 Sketch the current through dd'
- 4 Sketch the flux density at the location of $d' B(\pi/2, t)$. 5 Sketch the flux density at
- the location of $d' B(3\pi/2,t)$. 6 Sketch the torque on dd'.
- 7 Sketch the torque on qq'in the same way.





Results





Conclusions

- constant torque with two turns
- at synchronous speed: no torque
- torque proportional to rotor frequency (slip frequency) if the effect of the rotor currents on the field is negligible



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Induced voltage (also for SM)

• Air gap flux density:

$$B_{s}(\theta,t) = \hat{B}\cos(\theta - \omega t)$$

• Maximum air gap flux per pole (t=0):

$$\Phi_p = \iint \vec{B} \cdot \vec{n} \, \mathrm{d} \, A = \int_{-\pi/2}^{\pi/2} \hat{B} \cos(\theta - \omega t) lr \, \mathrm{d} \, \theta = 2lr \hat{B}$$

- Flux linkage of phase a if phase is a concentrated full-pitch winding with *N* turns: $\lambda_a = N\Phi_p \cos(\omega t)$
- Induced voltage:

$$e_a = \frac{\mathrm{d}\,\lambda_a}{\mathrm{d}\,t} = -\omega N\Phi_p \sin(\omega t)$$

In equation (5.23), a minus is inserted



Induced voltage and winding factor

$$\begin{cases} e_a = -\omega N \Phi_p \sin(\omega t) \\ e_b = -\omega N \Phi_p \sin(\omega t - \frac{2}{3}\pi) \\ e_c = -\omega N \Phi_p \sin(\omega t - \frac{4}{3}\pi) \end{cases}$$

$$E = \frac{\omega N \Phi_p}{\sqrt{2}} = \sqrt{2}\pi f N \Phi_p = 4.44 f N \Phi_p$$

A phase winding is generally not a concentrated full-pitch winding, but a distributed winding.

Therefore, the voltages induced in the different turns are not in phase and the resulting induced voltage is lower.

This is taken into account by including the winding factor.

$$E = \sqrt{2}k_w \pi N \Phi_p$$

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Induction machines

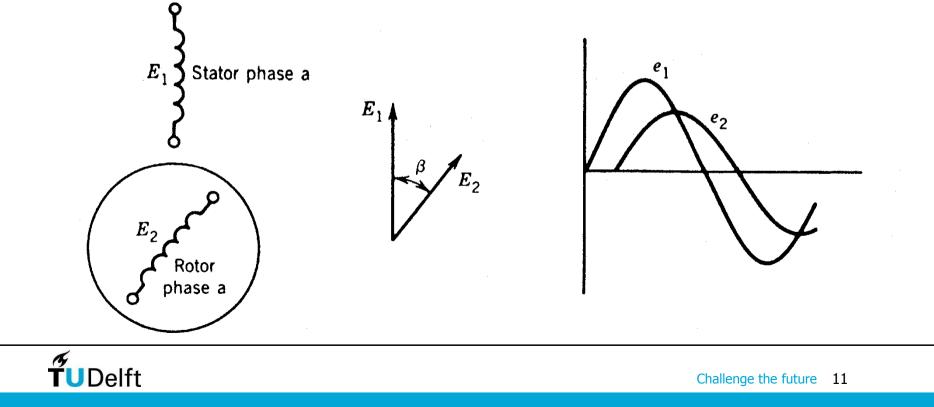
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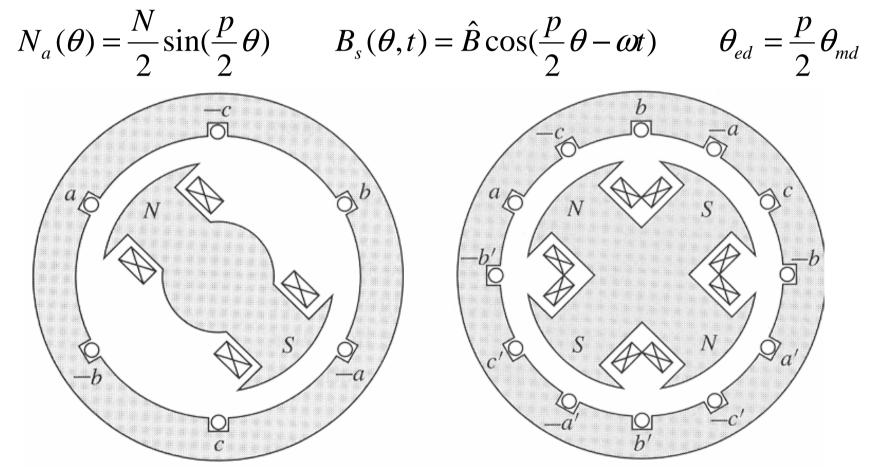
• A wound rotor induction machine with blocked rotor works as a transformer

$$E_1 = \sqrt{2}k_{w1}\pi f N_1 \Phi_p$$
$$E_2 = \sqrt{2}k_{w2}\pi f N_2 \Phi_p$$



p-pole machines

p is the number of poles (in some other books: number of pole-pairs)





Three-phase stator winding





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Running operation

 n_s is the synchronous speed of the rotor

$$n_s = f_1 60 \frac{2}{p}$$

for <i>f₁</i> =50Hz	p	2	4	6	8	10	20
	<i>n_s</i> (rpm)	3000	1500	1000	750	600	300

n is the actual speed of the rotor

s is the slip

$$s = \frac{n_s - n}{n_s} \qquad n = (1 - s)n_s \qquad n_s = n + sn_s$$

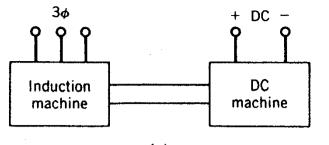
 f_2 is the rotor frequency $f_2 = sf_1$

$$E_{2s} = \frac{2\pi}{\sqrt{2}} f_2 k_{w2} N_2 \Phi_p = \frac{2\pi}{\sqrt{2}} s f_1 k_{w2} N_2 \Phi_p = s E_2$$

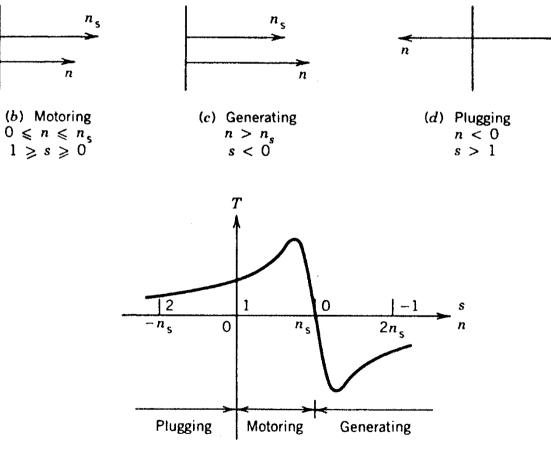
What is the speed of the rotor field with respect to the stator field?

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Three modes of operation







• demo

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 n_{s}

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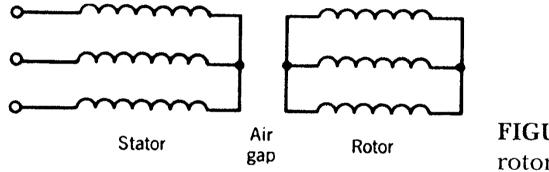
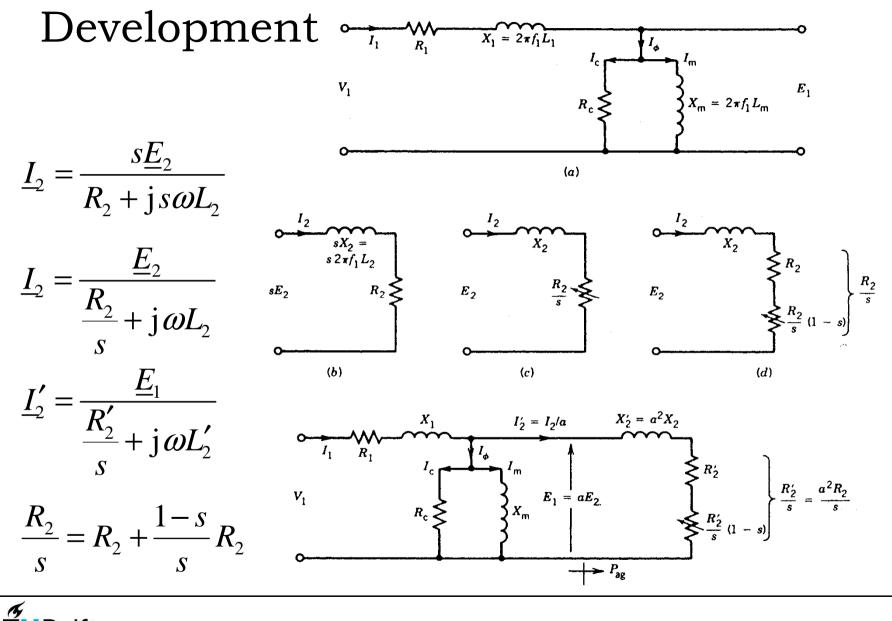


FIGURE 5.12 Three-phase wound rotor induction motor.

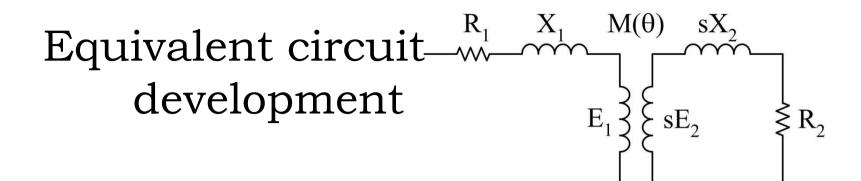
- When an induction machine is in standstill, it works as a transformer.
- When the rotor rotates, also a voltage will be induced in the rotor windings.
- Therefore an equivalent circuit is assumed which is comparable to a transformer.



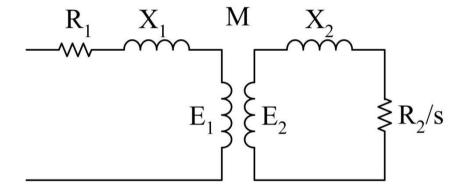


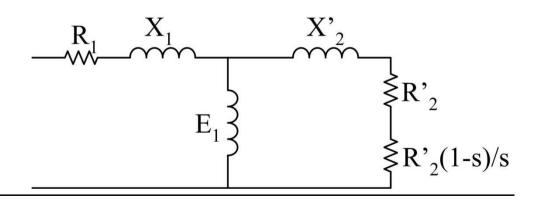
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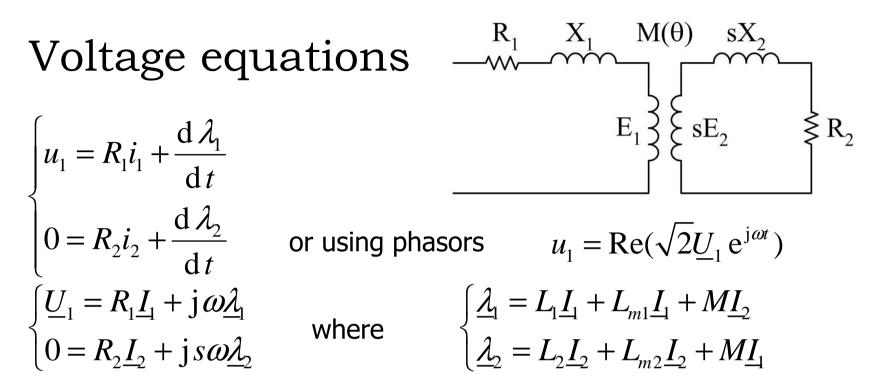


- L₁ and L₂ are the stator and rotor leakage inductances.
- Where are the leakage fields?









 L_1 and L_2 are the stator and rotor leakage inductances L_{m1} and L_{m2} and are the stator and rotor main inductances that give the ratio between the phase flux linkage and the phase current assuming three-phase currents

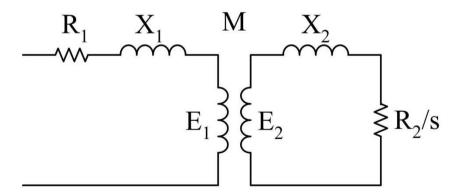
M is the mutual inductance between stator and rotor that gives the ratio between the phase flux linkage and the phase current assuming three-phase currents



Transformations of the voltage equations

The rotor voltage equation is divided by the slip:

$$\begin{cases} \underline{U}_{1} = R_{1}\underline{I}_{1} + j\omega L_{1}\underline{I}_{1} + j\omega L_{m1}\underline{I}_{1} + j\omega M\underline{I}_{2} = R_{1}\underline{I}_{1} + j\omega L_{1}\underline{I}_{1} + \underline{E}_{1} \\ 0 = \frac{R_{2}}{s}\underline{I}_{2} + j\omega L_{2}\underline{I}_{2} + j\omega L_{m2}\underline{I}_{2} + j\omega M\underline{I}_{1} = \frac{R_{2}}{s}\underline{I}_{2} + j\omega L_{2}\underline{I}_{2} + \underline{E}_{2} \end{cases}$$





Transformations of voltage equations

Referring the rotor quantities to the stator in a power invariant way

$$\underline{E}_{2}' = \frac{L_{m1}}{M} \underline{E}_{2} \qquad \underline{I}_{2}' = \frac{M}{L_{m1}} \underline{I}_{2} \qquad R_{2}' = \frac{L_{m1}^{2}}{M^{2}} R_{2} \qquad L_{2}' = \frac{L_{m1}^{2}}{M^{2}} L_{2}$$
using $M^{2} = L_{m1}L_{m2}$ gives
$$\begin{cases} \underline{U}_{1} = R_{1}\underline{I}_{1} + j\omega L_{1}\underline{I}_{1} + j\omega L_{m1}(\underline{I}_{1} + \underline{I}_{2}') = R_{1}\underline{I}_{1} + j\omega L_{1}\underline{I}_{1} + \underline{E}_{1} \\ 0 = \frac{R_{2}'}{s} \underline{I}_{2}' + j\omega L_{2}'\underline{I}_{2}' + j\omega L_{m1}(\underline{I}_{1} + \underline{I}_{2}') = \frac{R_{2}'}{s} \underline{I}_{2}' + j\omega L_{2}'\underline{I}_{2}' + \underline{E}_{1} \\ \hline R_{1}' \qquad R_{1}'' \qquad R_{2}'' \\ \hline R_{1}'' \qquad R_{2}'' \qquad R_{2}'' \\ \hline R_{1}'' \qquad R_{2}'' \\ \hline R_{1}'' \qquad R_{2}'' \\ \hline R_{2}''$$

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Γ-equivalent circuit

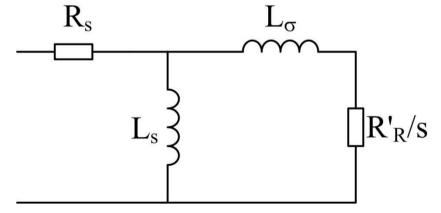
$$\begin{cases} \underline{U}_{1} = R_{1}\underline{I}_{1} + j\omega L_{1}\underline{I}_{1} + j\omega L_{m1}(\underline{I}_{1} + \underline{I}_{2}') \\ 0 = \frac{R_{2}'}{s}\underline{I}_{2}' + j\omega L_{2}'\underline{I}_{2}' + j\omega L_{m1}(\underline{I}_{1} + \underline{I}_{2}') \end{cases} \quad \text{substituting} \quad \begin{array}{l} \underline{I}_{2}' = \frac{L_{s}}{L_{m1}}\underline{I}_{R}' \\ L_{s} = L_{m1} + L_{1} \end{cases}$$

$$\begin{cases} \underline{U}_{1} = R_{1}\underline{I}_{1} + j\omega L_{s}(\underline{I}_{1} + \underline{I}_{R}') \\ 0 = \frac{R_{R}'}{s}\underline{I}_{2}' + j\omega L_{\sigma}\underline{I}_{R}' + j\omega L_{s}(\underline{I}_{1} + \underline{I}_{R}') \end{cases}$$

 $|L_s|$

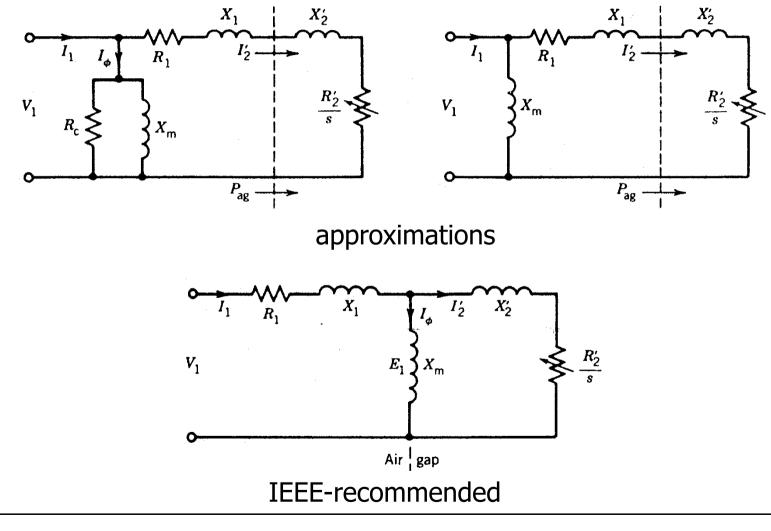
where

$$L_{\sigma} = \left(\frac{L_{s}(L_{m1} + L_{2})}{L_{m1}^{2}} - 1\right)$$
$$R_{R}' = \frac{L_{s}^{2}}{L_{m1}^{2}}R_{2}'$$



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Other equivalent circuits





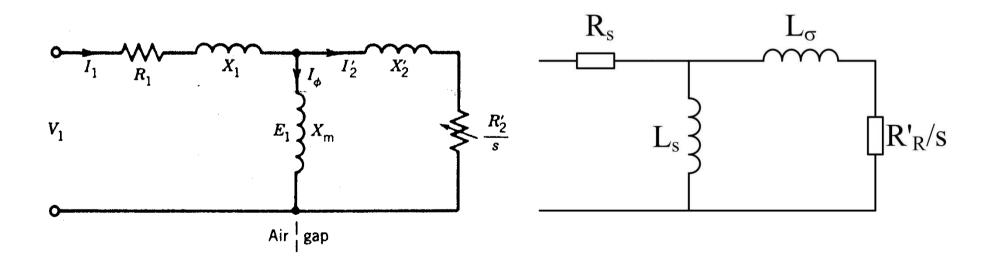
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Parameter identification (5.8)

- resistance measurement (dc current) which resistance?
- no-load test which parameters?
- blocked-rotor test which parameters?



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