

# Equation Handout

WB2301 System Identification & Parameter Estimation

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## 1 Time domain

### 1.1 Time signals

$$u(i), y(i); \quad i \in [1, 2, 3, \dots, N]$$
$$\Delta t = \frac{1}{f_s} = \frac{T}{N}$$

### 1.2 Cross-correlation function

$$\hat{\phi}_{uy}(\tau) = \frac{1}{N} \sum_{i=\tau}^N u(i-\tau)y(i); \quad |\tau| < N \quad (1)$$

$$= \text{xcorr}(\mathbf{y}, \mathbf{u}, 'biased') \quad (2)$$

The cross-product function is scaled by  $N$ . Division by  $N$  gives the 'biased' estimator which is preferred over the 'unbiased' estimator (division by  $N - |\tau|$ ).

### 1.3 Cross-covariance function

$$C_{uy}(\tau) = \frac{1}{N} \sum_{i=\tau}^N (u(i-\tau) - \mu_u)(y(i) - \mu_y); \quad |\tau| < N \quad (3)$$

$$= \text{xcov}(\mathbf{y}, \mathbf{u}, 'biased') \quad (4)$$

$$C_{uu}(0) = \sigma_u^2 \quad (5)$$

The righthand-side of (3) can be expanded as:

$$\frac{1}{N} \sum_{i=\tau}^N (u(i-\tau)y(i) - \frac{1}{N} \sum_{i=\tau}^N u(i-\tau)\mu_y - \frac{1}{N} \sum_{i=\tau}^N y(i)\mu_u + \frac{1}{N} \sum_{i=\tau}^N \mu_u\mu_y) \quad (6)$$

The first term of (6) equals  $\hat{\phi}_{uy}$  (Eq. 1). For a large number of summation elements ( $N - \tau \approx N$ ), each of the last three terms converge to  $\mu_u\mu_y$  and Eq. (7) can be used. Note that in case an 'unbiased' estimator is used the fourth term of (6) equals  $\mu_u\mu_y$  straightforward.

$$C_{uy}(\tau) = \hat{\phi}_{uy}(\tau) - \mu_u\mu_y \quad N - \tau \approx N \quad (7)$$

## 1.4 Cross-correlation function

$$\begin{aligned}r_{uy}(\tau) &= \frac{1}{N} \sum_{i=\tau}^N \frac{(u(i-\tau) - \mu_u)(y(i) - \mu_y)}{\sigma_u \sigma_y}; \quad |\tau| < N \\ &= \frac{C_{uy}(\tau)}{\sigma_u \sigma_y} \\ &= \text{xcov}(y, u, 'coeff')\end{aligned}$$

## 2 Frequency domain

### 2.1 Discrete Fourier Transform (DFT)

$$\begin{aligned}U(f) &= \text{DFT}\{u(i)\} = \sum_{i=1}^N u(i)e^{-j2\pi \frac{fi}{N}}; f = 0, \dots, N-1 \\ &= \text{fft}(u) \\ u(i) &= \text{DFT}^{-1}\{U(f)\} = \frac{1}{N} \sum_{f=0}^{N-1} U(f)e^{j2\pi \frac{fi}{N}}; i = 1, \dots, N \\ &= \text{ifft}(U)\end{aligned}$$

### 2.2 Frequency signals

$$\begin{aligned}U(f), Y(f); \quad f \in [0, 1, 2, \dots, N-1] \\ \Delta f = \frac{1}{N\Delta t} = \frac{1}{T} \\ U(-f) = U^*(f)\end{aligned}$$

Note that the  $U(f)$  is mirrored and only contains relevant information for  $f \in [0, 1, 2, \dots, \frac{N}{2}]$ .

### 2.3 Spectral densities via cross-covariance function

$$S_{uy} = \text{DFT}\{C_{uy}\}$$

### 2.4 Spectral densities via Fourier transformed signals

$$\begin{aligned}S_{uu}(f) &= \frac{1}{N}U(f)U(-f) = \frac{1}{N}U(f)U^*(f) = \frac{1}{N}|U(f)|^2 \\ &= \frac{1}{N}|\text{fft}(u)|^2 \\ S_{uy}(f) &= \frac{1}{N}Y(f)U(-f) = \frac{1}{N}Y(f)U^*(f)\end{aligned}$$

## 2.5 Parseval's Theorem

$$RMS_u = \sqrt{\frac{\sum u^2}{N}}$$

Parseval's theorem gives the relation between the Root-Mean-Squared (RMS) of a signal and the integral of the auto spectral density:

$$\begin{aligned} RMS_u^2 &= \frac{1}{N} \sum_{f=0}^{N-1} S_{uu}(f) \\ \frac{\sum u^2}{N} &= \frac{1}{N} \sum_{f=0}^{N-1} \frac{1}{N} |U(f)|^2 = \frac{1}{N^2} \sum_{f=0}^{N-1} |U(f)|^2 \\ \frac{\sum u^2}{N} &= \frac{1}{N^2} \sum |\text{fft}(\mathbf{u})|^2 \end{aligned}$$

Note that in case of a zero-mean signal the RMS value (approximately) equals the standard deviation, i.e. the integral of the auto-spectral density equals the signal variance:

$$C_{uu}(0) = \sigma_u^2 = \frac{1}{N} \sum_{f=0}^{N-1} S_{uu}(f) = \frac{2}{N} \sum_{f=0}^{\frac{N}{2}-1} S_{uu}(f); \quad \text{if } \mu_u = 0$$

## 3 Identification of transfer functions

### 3.1 Open loop: time domain approach

$$\begin{aligned} y(t) &= n(t) + \int g(t')u(t-t')dt' \\ y(t+\tau)u(t) &= n(t+\tau)u(t) + \int g(t')u(t+\tau-t')u(t)dt' \\ C_{yy}(\tau) &= C_{nu}(\tau) + \int g(t')C_{uu}(\tau)dt' \end{aligned}$$

if  $u(t)$  is white noise ( $C_{uu}(\tau) = \sigma_u^2$  if  $\tau = 0$  and  $C_{uu}(\tau) = 0$  if  $\tau \neq 0$ ) then holds:

$$C_{yy}(\tau) = C_{nu}(\tau) + g(\tau)\sigma_u^2$$

### 3.2 Open loop: frequency domain approach

Indirect approach (Fourier transform of covariances)

$$\begin{aligned} C_{yy}(\tau) &= C_{nu}(\tau) + \int g(t')C_{uu}(\tau)dt' \\ S_{yy}(f) &= S_{nu}(f) + G(f)S_{uu}(f) \end{aligned}$$

Direct approach (Fourier transform of signals)

$$Y(f) = G(f)U(f) + N(f)$$

$$Y(f)U^*(f) = G(f)U(f)U^*(f) + N(f)U^*(f)$$

$$S_{yy} = G(f)S_{uu}(f) + S_{nu}(f)$$

if  $u$  and  $n$  are not correlated ( $S_{nu} = 0$ ):

$$H(f) = \frac{S_{yy}(f)}{S_{uu}(f)}; \quad \text{if } S_{nu}(f) = 0 \forall f$$