

# System- en Regeltechniek II

Lecture 10 – Nyquist plot and stability criterium

Robert Babuška

*Delft Center for Systems and Control  
Faculty of Mechanical Engineering  
Delft University of Technology  
The Netherlands*

e-mail: r.babuska@dcsc.tudelft.nl  
www.dcsc.tudelft.nl/~babuska  
tel: 015-27 85117

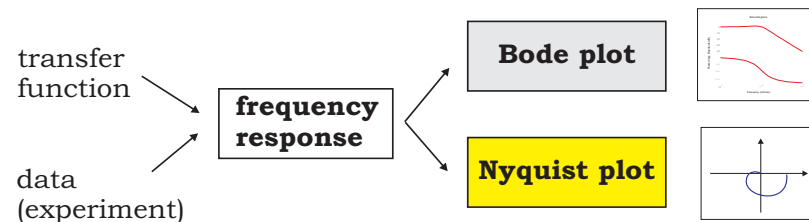
## Lecture Outline

Previous lecture: PID controller design, lead and lag compensators.

Today:

- Nyquist plot.
- Nyquist stability criterion.

## Frequency Domain Methods



## Frequency Domain Methods

The key of frequency domain design:

provide sufficient phase at the crossover frequency

(= get the closed-loop far enough from the point of becoming unstable)

⇒ Bode plots are well suited as a design and analysis tool.

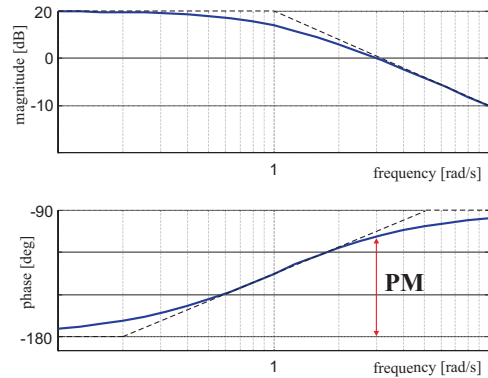
... , so, do we need yet another kind plot?

In fact, we do, let's have a look why ...

## Motivating Example

$$G(s) = \frac{10}{s-1}$$

sketch the bode plot, indicate PM



## Deficiency of Bode Plots

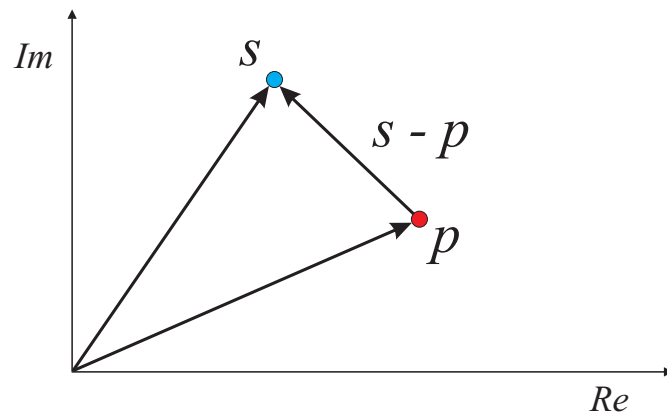
For systems with poles in right half-plane, the Bode plot alone does not provide any good indication of stability / instability.

→ In the above example, the phase will never cross  $-180^\circ$ , and yet, for  $K < 0.1$ , the closed loop becomes unstable (check the root locus!).

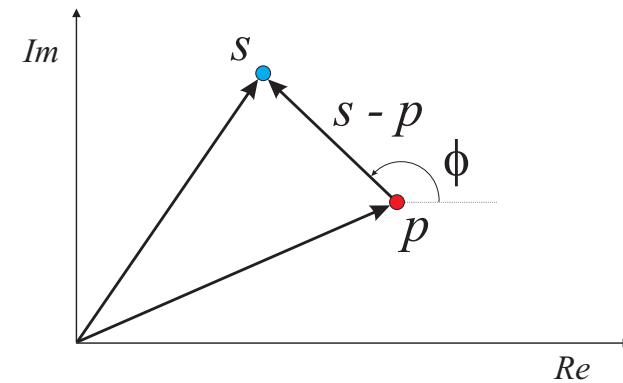
Is there a method for frequency domain design, considering stability for all kinds of systems?

Yes, the Nyquist plot and stability criterion.

## Complex Numbers as Vectors



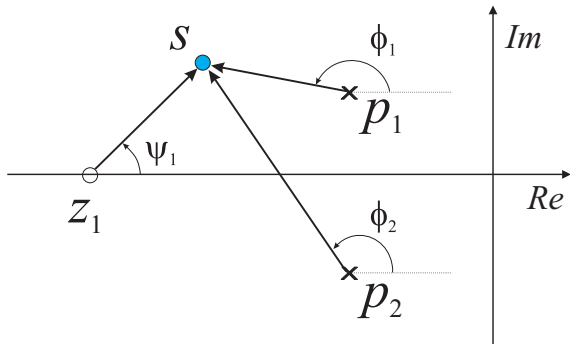
## Euler Representation



$$s - p = |s - p| e^{j\phi}$$

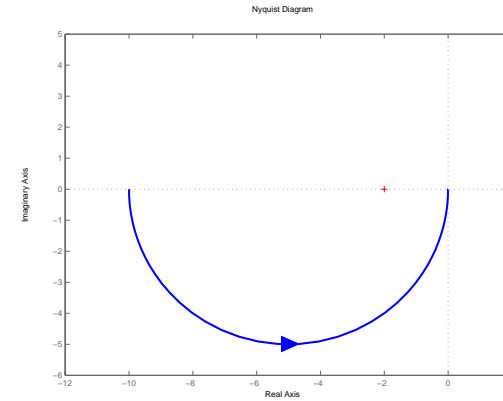
## Transfer Function

$$G(s) = \frac{s - z_1}{(s - p_1)(s - p_2)} = \frac{|s - z_1|}{|s - p_1| \cdot |s - p_2|} e^{j(\psi_1 - \phi_1 - \phi_2)}$$



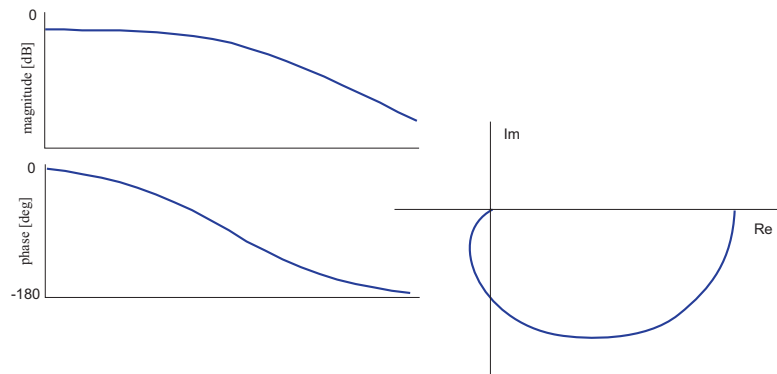
## Nyquist (Polar) Plot

Let  $s = j\omega$  for  $\omega \in [0, \infty)$  and plot  $\text{Im}[G(j\omega)]$  against  $\text{Re}[G(j\omega)]$ .

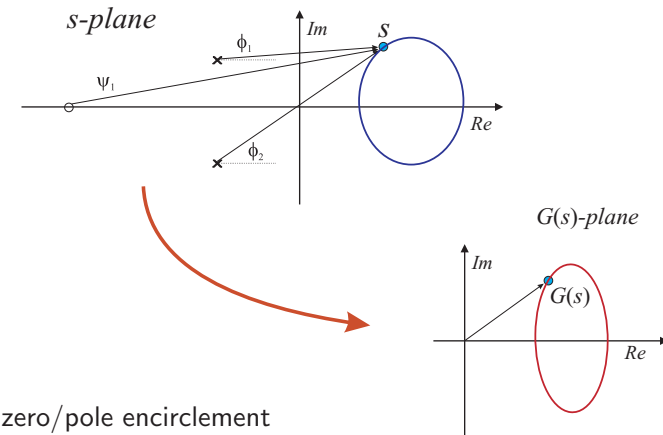


## Relation Bode Plot – Nyquist Plot

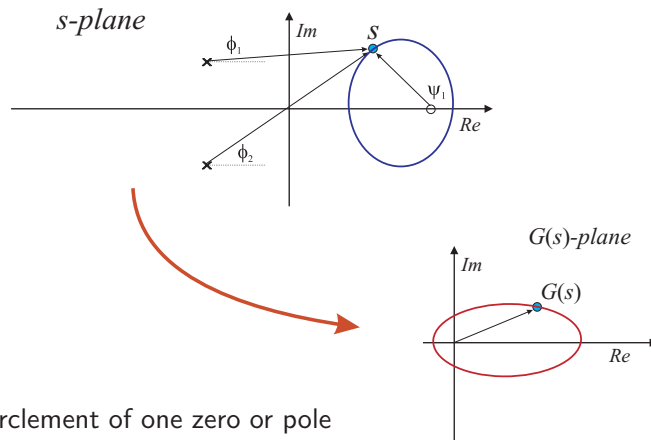
$$G(s) = \frac{1}{(s + 1)(s + 5)}$$



## Argument Principle (I)



## Argument Principle (II)



Encirclement of one zero or pole  
 $\implies$  one encirclement of the origin

## Argument Principle in General

For a clockwise contour in the  $s$ -plane, denote:

$P$  number of poles encircled in the  $s$ -plane

$Z$  number of zeros encircled in the  $s$ -plane

$N$  number of clockwise encirclements of the origin by  $G(s)$

$$N = Z - P$$

Recall:

$$\angle G(s) = \sum_i \angle(s - z_i) - \sum_j \angle(s - p_j) = \sum_i \psi_i - \sum_j \phi_j$$

## Argument Principle for Stability Analysis

Given the Nyquist plot of  $KG(s)$ , we want to determine whether the closed loop:

$$G_{cl}(s) = \frac{Y(s)}{R(s)} = \frac{KG(s)}{1 + KG(s)}$$

is stable.

- Closed-loop stability  $\iff G_{cl}(s)$  has no poles in RHP.
- Poles are given by  $1 + KG(s)$ , so let us study  $1 + KG(s)$ .

## Argument Principle for Stability Analysis

$$G_{cl}(s) = \frac{Y(s)}{R(s)} = \frac{KG(s)}{1 + KG(s)}$$

Poles of  $G_{cl}(s)$  are the solutions of  $1 + KG(s) = 0$ , i.e.:

poles of  $G_{cl}(s)$  are the zeros of  $(1 + KG(s))$

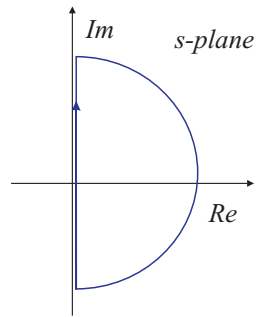
in addition, as:

$$1 + KG(s) = 0 \rightarrow 1 + K \frac{b(s)}{a(s)} = 0 \rightarrow \frac{a(s) + Kb(s)}{a(s)} = 0$$

poles of  $G(s)$  are the poles of  $(1 + KG(s))$

## Argument Principle for Stability Analysis

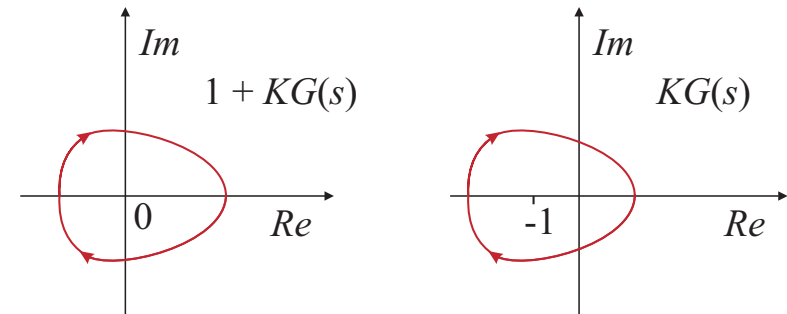
So, we want to find out, if  $1 + KG(s)$  has no RHP zeros.



encircle the entire RHP

= draw the Nyquist diagram for frequencies  $\omega \in (-\infty, \infty)$

## Argument Principle for Stability Analysis



Draw the Nyquist diagram of the loop TF  $L(s) = KG(s)$ ,  
count clockwise encirclements of  $-1$  :  $N = Z - P$

## Argument Principle for Stability Analysis

$$N = Z - P$$

$Z$  = number of RHP zeros of  $(1 + KG(s))$

$P$  = number of RHP poles of  $(1 + KG(s))$

Given that:

$$1 + KG(s) = 0 \rightarrow \frac{a(s) + Kb(s)}{a(s)} = 0$$

we have:

$Z$  = number of RHP poles of  $G_{cl}(s)$  ... CL poles

$P$  = number of RHP poles of  $G(s)$  ... OL poles

## Nyquist Stability Criterion

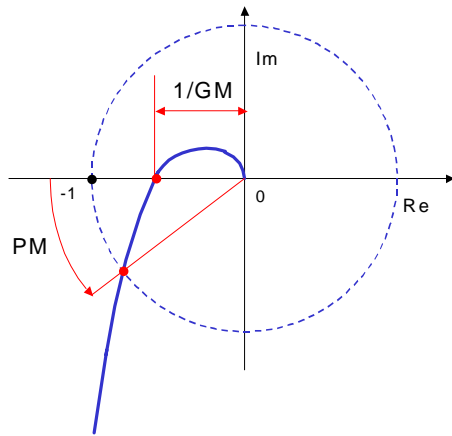
$$Z = N + P$$

In words:

number of RHP closed-loop poles =

clockwise encirclements + number of RHP open-loop poles

## Stability Margins in Nyquist Plot



## Why Don't We Ride These Bikes?



## Simple Bicycle Models

Front steering:

$$G(s) = \frac{\phi(s)}{\delta(s)} = K \frac{s + \frac{v}{a}}{s^2 - \frac{q}{h}}$$

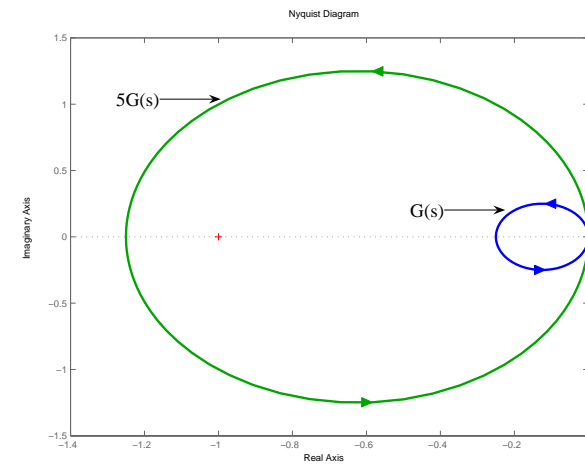
Rear steering:

$$G(s) = \frac{\phi(s)}{\delta(s)} = K \frac{-s + \frac{v}{a}}{s^2 - \frac{q}{h}}$$

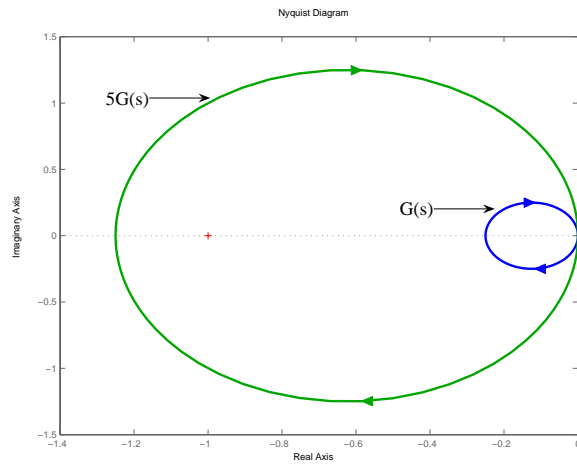
$a$  – distance of COM to fixed wheel center

$h$  – height of COM above ground

## Front Steering: Nyquist Plot



## Rear Steering: Nyquist Plot



## Nyquist: Homework Assignments

- Read Section 6.3 (Nyquist stability criterion).
- Work out examples in this section and verify the results by using Matlab.
- Work out problems 6.18 – 6.22 and verify your results by using Matlab.